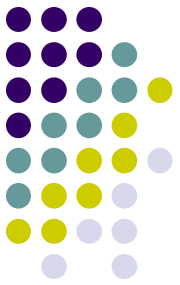


Drift and Diffusion Currents

- **Current**
Generated by the movement of charged particles (negatively charged electrons and positively charged holes).
- **Carriers**
The charged electrons and holes are referred to as **carriers**
- The two basic processes which cause electrons and holes move in a semiconductor:
 - **Drift** - the movement caused by electric field.
 - **Diffusion** - the flow caused by variations in the concentration.



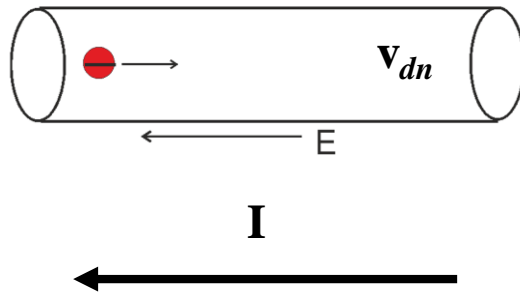


Drift Currents

- **Drift Current Density**

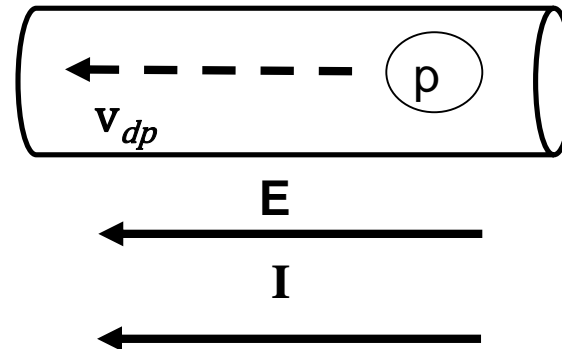
- When an electric field E is applied to a semiconductor, it will produce a force that causes the carriers to move

N-type



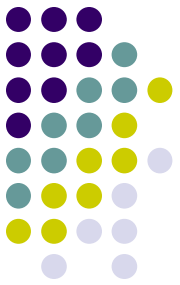
- Produces a force on the electrons in the **opposite** direction of the Electric Field, because of the electrons' negative charge.
- The electrons acquire a drift velocity, v_{dn} (in cm/s):

P-type



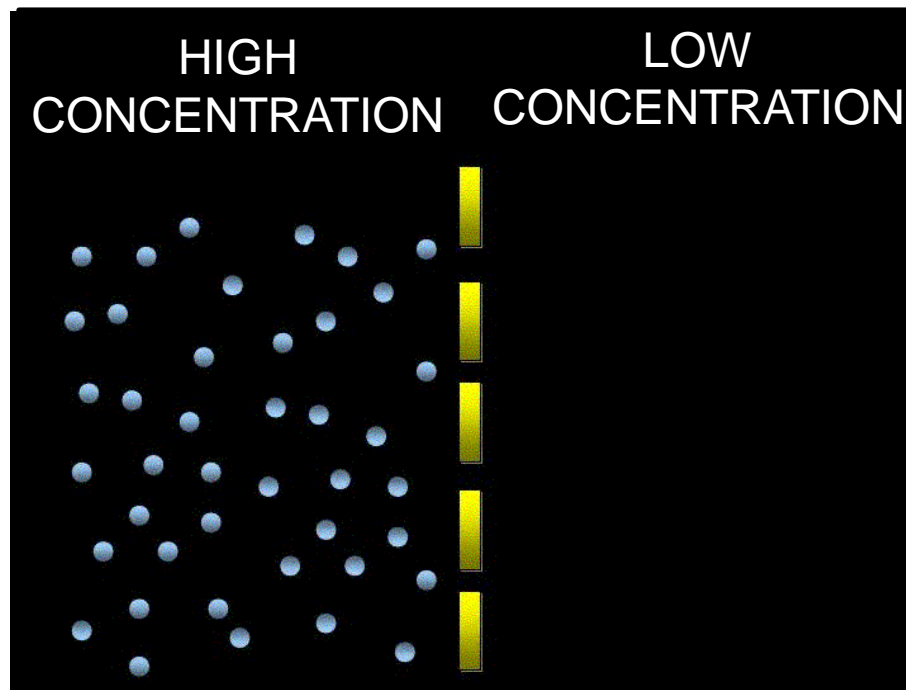
- Produces a force on the holes in the **same** direction of the Electric Field, because of the positive charge on the holes.
- The holes acquire a drift velocity, v_{dp} (in cm/s):



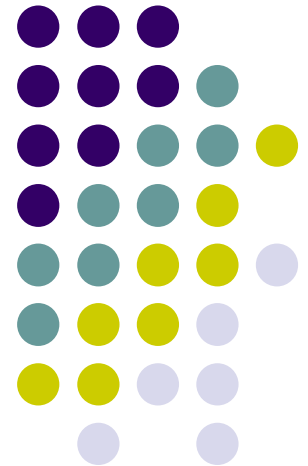


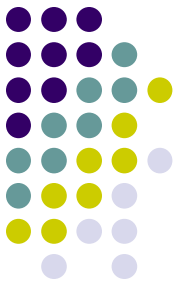
Diffusion Current

- The basic diffusion process
 - Flow of particles **from** a region of **high**-concentration to a region of **low**-concentration.
 - The movement of the particles will then generate the diffusion current



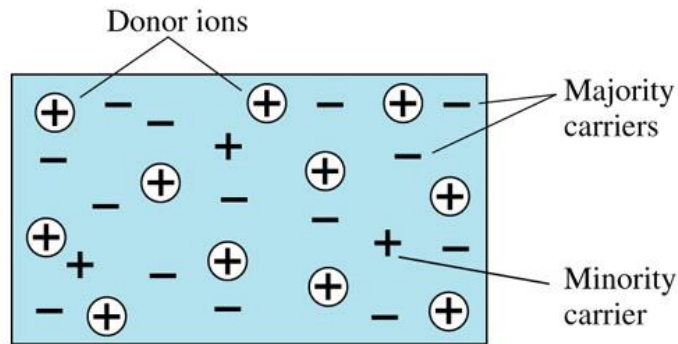
The PN Junction





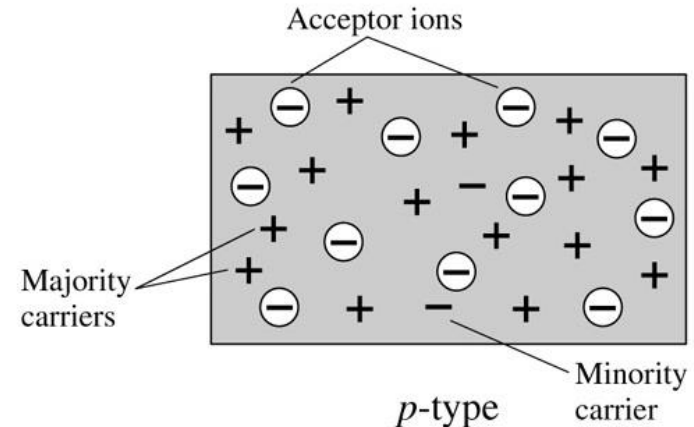
n-type versus p-type

- In n-type - the electrons are the majority carriers and holes are the minority carriers.
- In p-type - the holes are called the majority carriers and electrons are the minority carriers.



n-type

(a)

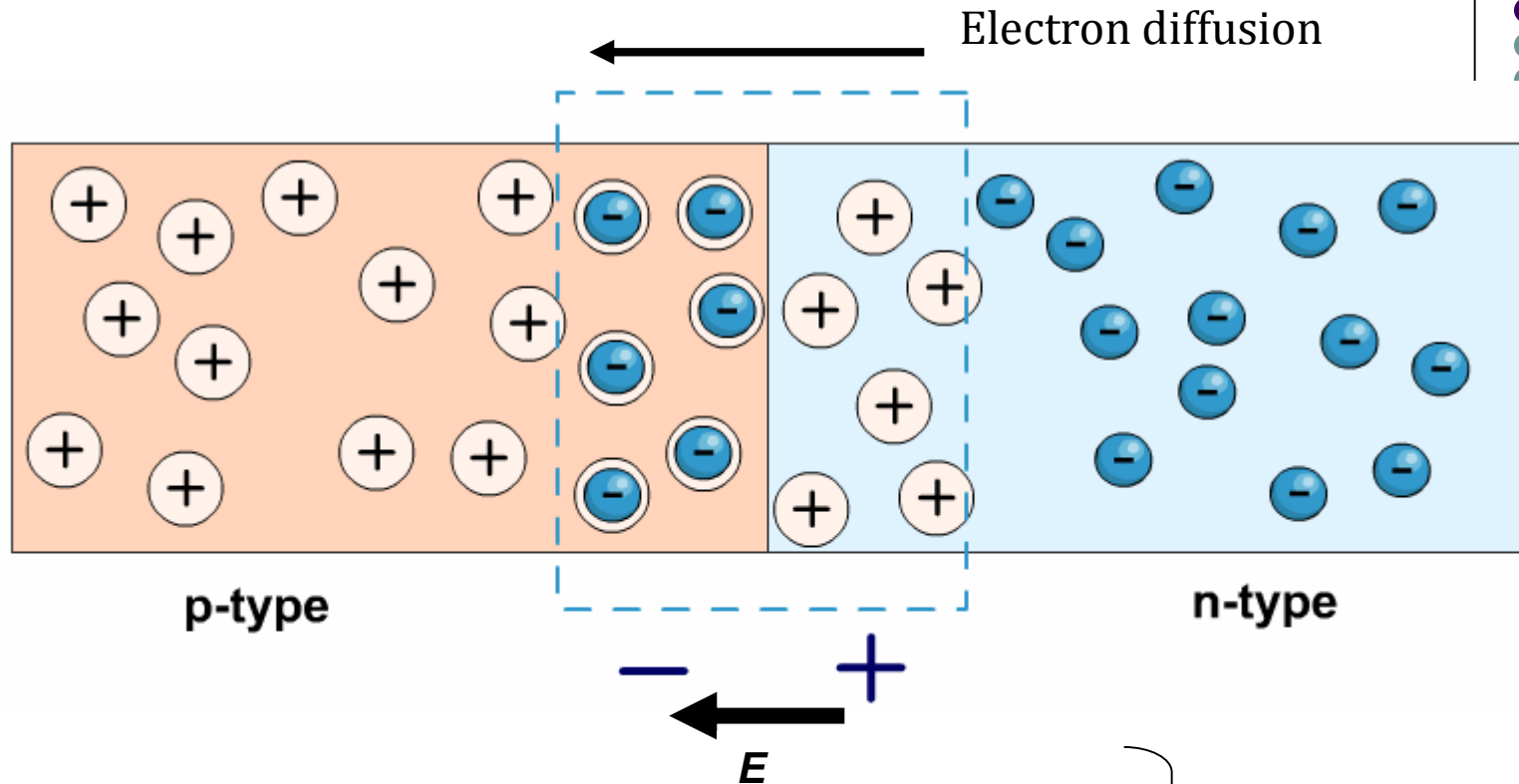


p-type

(b)



The Equilibrium pn Junction



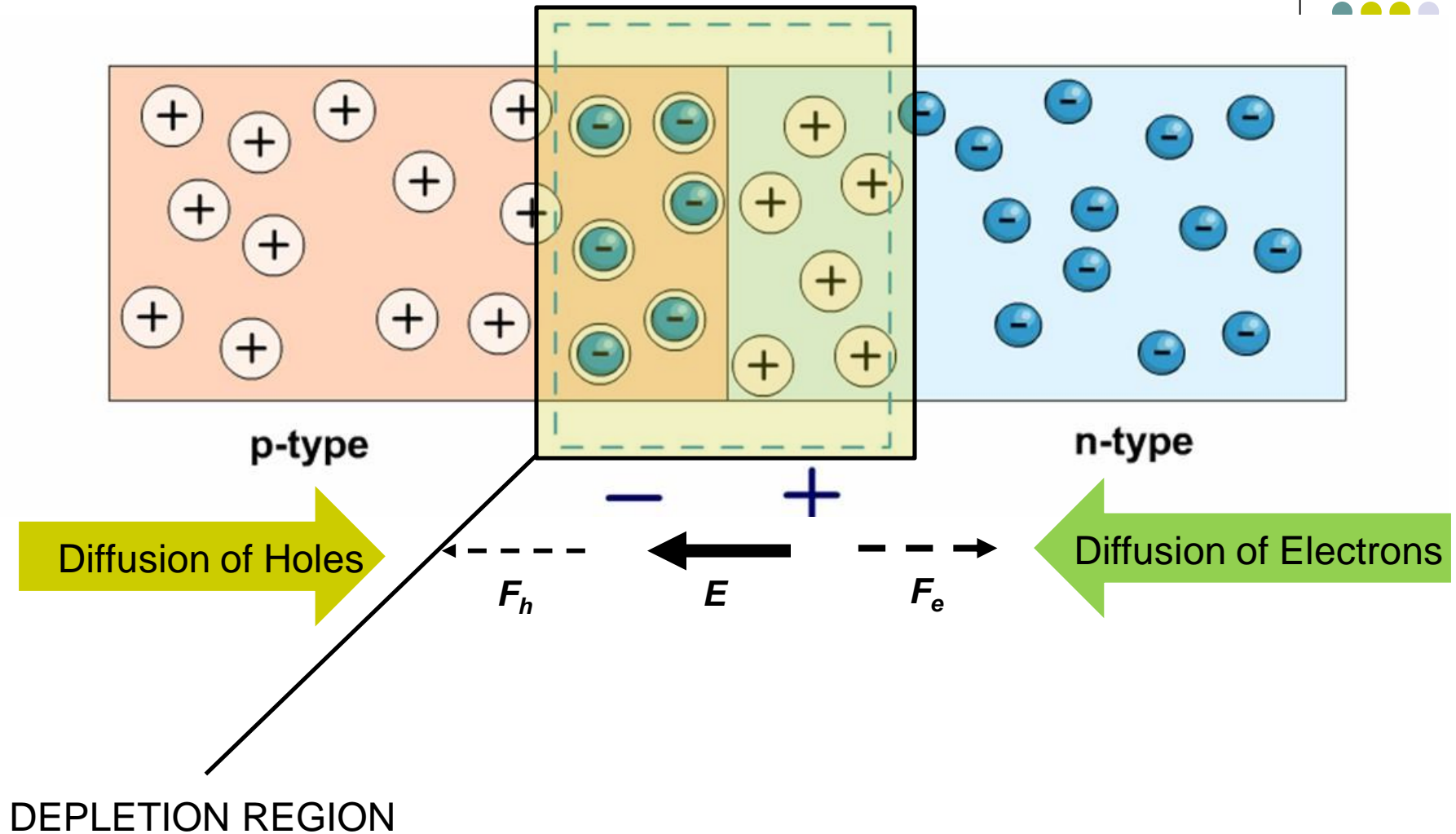
Positive and negative ions are formed by gaining or losing electrons.

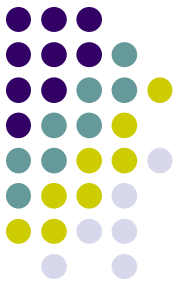
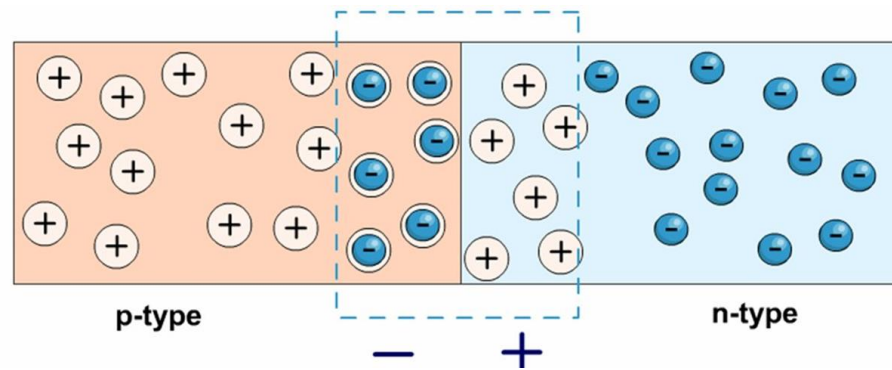
- **positively** charged **ions** by losing electrons
- **negatively** charged **ions** by gaining electrons.

Creates a charge separation that produces an electric field, E



The Electric field will create a force that will stop the diffusion of carriers \rightarrow reaches thermal equilibrium condition





Known as space charge region/depletion region.

Potential difference across the depletion region is called the built-in potential barrier, or built-in voltage:

$$V_{bi} = \frac{kT}{e} \ln\left(\frac{N_a N_d}{n_i^2}\right) = V_T \ln\left(\frac{N_a N_d}{n_i^2}\right)$$

V_T = thermal voltage = kT/e

k = Boltzmann's constant = 86×10^{-6} eV/K = 1.38×10^{-23} J/K

T = absolute temperature

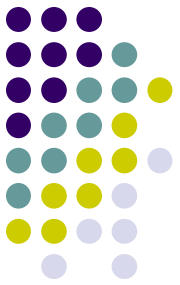
e = the magnitude of the electron charge = 1.6×10^{-19} C

N_a = the net acceptor concentration in the p-region

N_d = the net donor concentration in the n-region

NOTE: V_T = **thermal voltage**, [$V_T = kT / e$] it is approximately **0.026 V** at temp, $T = 300$ K





The Equilibrium pn Junction

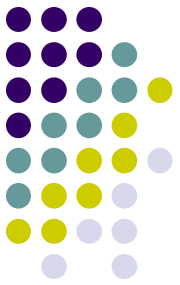
Example 1

Calculate the built-in potential barrier of a pn junction. Consider a silicon pn junction at $T = 300$ K, doped $N_a = 10^{16} \text{ cm}^{-3}$ in the p-region, $N_d = 10^{17} \text{ cm}^{-3}$ in the n-region and $n_i = 1.5 \times 10^{10} \text{ cm}^{-3}$.

Solution

$$V_{bi} = V_T \ln\left(\frac{N_a N_d}{n_i^2}\right) = (0.026) \ln\left[\frac{(10^{16})(10^{17})}{(1.5 \times 10^{10})^2}\right] = 0.757 \text{ V}$$



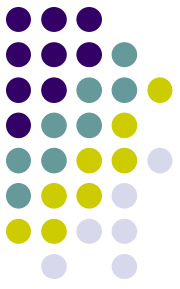


$$V_{bi} = \frac{kT}{e} \ln\left(\frac{N_a N_d}{n_i^2}\right) = V_T \ln\left(\frac{N_a N_d}{n_i^2}\right)$$

- Example 2

Consider a silicon pn junction at $T = 400\text{K}$, doped with concentrations of $N_d = 10^{18} \text{ cm}^{-3}$ in n-region and $N_a = 10^{19} \text{ cm}^{-3}$ in p-region. **Calculate the built-in voltage V_{bi} of the pn junction**, given B and E_g for silicon are $5.23 \times 10^{15} \text{ cm}^{-3} \text{ K}^{-3/2}$ and 1.1 eV respectively





Consider a silicon pn junction at $T = 400\text{K}$, doped with concentrations of $N_d = 10^{18} \text{ cm}^{-3}$ in n-region and $N_a = 10^{19} \text{ cm}^{-3}$ in p-region. **Calculate the built-in voltage V_{bi} of the pn junction**, given B and E_g for silicon are $5.23 \times 10^{15} \text{ cm}^{-3} \text{ K}^{-3/2}$ and 1.1 eV respectively

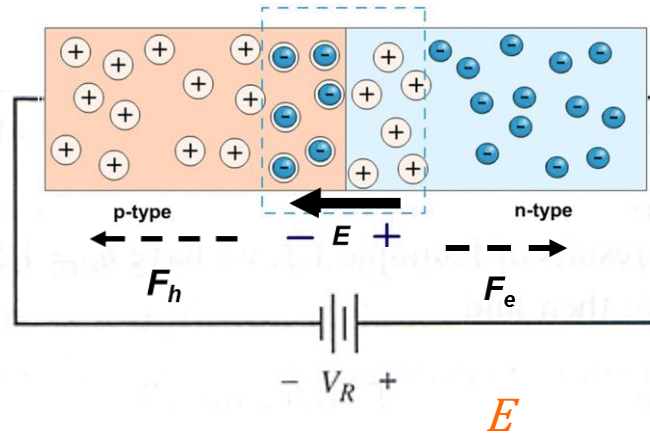
- Calculation of $V_T =$
 $=$
- Calculation of $n_i = BT^{3/2} \exp (-E_g / 2kT)$
- Calculation of $V_{bi} = V_T \ln (N_a N_d / n_i^2)$



Reverse-Biased pn Junction



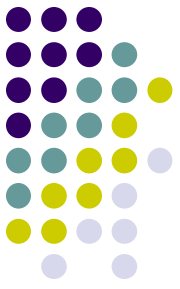
- +ve terminal is applied to the n-region of the pn junction and vice versa.



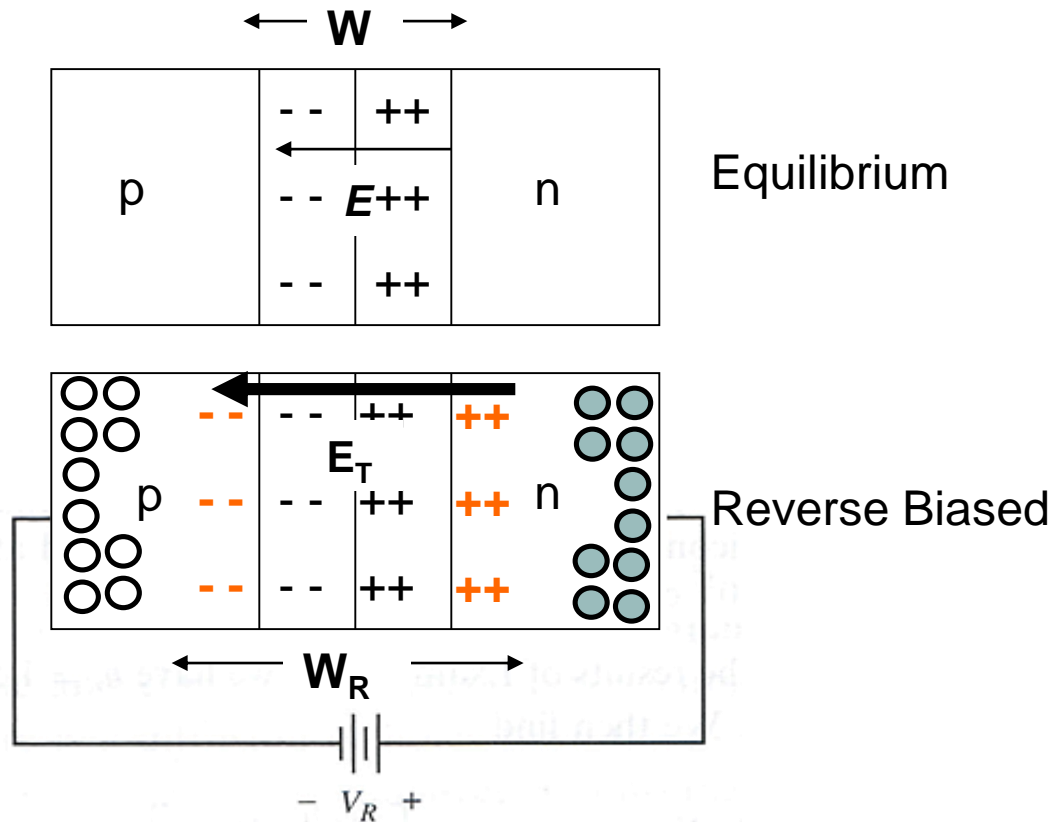
- Applied voltage V_R will induce an applied electric field E_A .
- Direction of the E_A is the same as that of the E -field in the space-charge region.
- Magnitude of the electric field in the space-charge region increases above the thermal equilibrium value. Total $E_T = E + E_A$
- Increased electric field holds back the holes in the p-region and the electrons in the n-region.
- Hence, no current will flow through the pn junction except for a very small reverse bias current that can be neglected

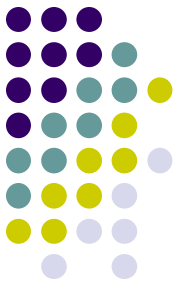


Reverse-Biased pn Junction



- The depletion region is wider during reverse biased
- The negative terminal of the power supply attracts the holes from the p-region and the positive terminal attract the electrons from p-region.
- So, the majority charge carriers move away from the junction. This increases the width of the depletion layer.





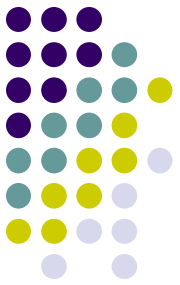
The space charge region/depletion region increase with increase of reverse-bias voltage, so a capacitor is associated with the pn junction when a reverse-bias voltage is applied. The junction capacitance or depletion layer capacitance of a pn junction is

$$C_j = C_{j0} \left(1 + \frac{V_R}{V_{bi}} \right)^{-1/2}$$

EXAMPLE 2.4 Calculate the junction capacitance of a silicon pn junction diode. Consider that the diode is at room temperature ($T = 300^\circ\text{K}$), with doping concentrations of $N_a = 1.5 \times 10^{16} \text{ cm}^{-3}$, $N_d = 1.0 \times 10^{15} \text{ cm}^{-3}$ and let $C_{j0} = 1.5 \text{ pF}$. Calculate the junction capacitance at reverse bias 3.5 V.



$$V_{bi} = \frac{kT}{e} \ln\left(\frac{N_a N_d}{n_i^2}\right) = V_T \ln\left(\frac{N_a N_d}{n_i^2}\right)$$



$$V_b = (0.026) \ln\left[\frac{(1.5 \times 10^{16})(1.0 \times 10^{15})}{(1.5 \times 10^{10})^2}\right] = 0.6479 \text{ V}$$

The junction capacitance at $V_R = 3.5 \text{ V}$ is

$$C_j = C_{j0} \left(1 + \frac{V_R}{V_b}\right)^{-1/2} = (1.5) \left(1 + \frac{3.5}{0.6479}\right)^{-1/2} = 0.5928 \text{ pF}$$

less capacitance



more capacitance



$$C = \frac{\epsilon A}{d}$$

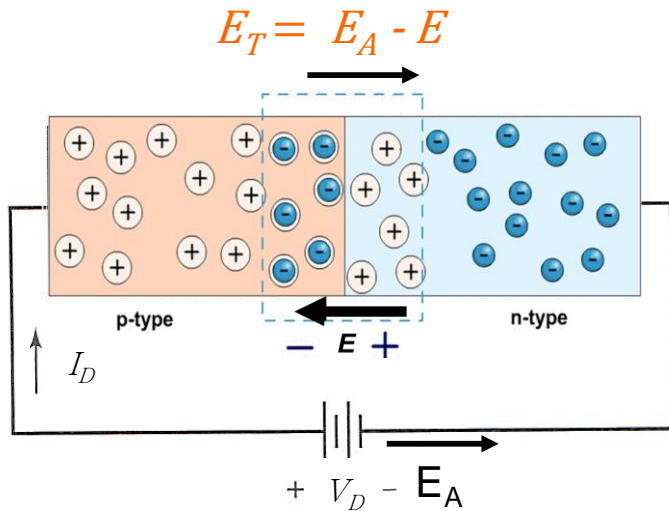
Width of depletion region increases



Before applying reverse voltage, C_{j0}

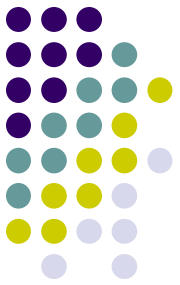


Forward-Biased pn Junction

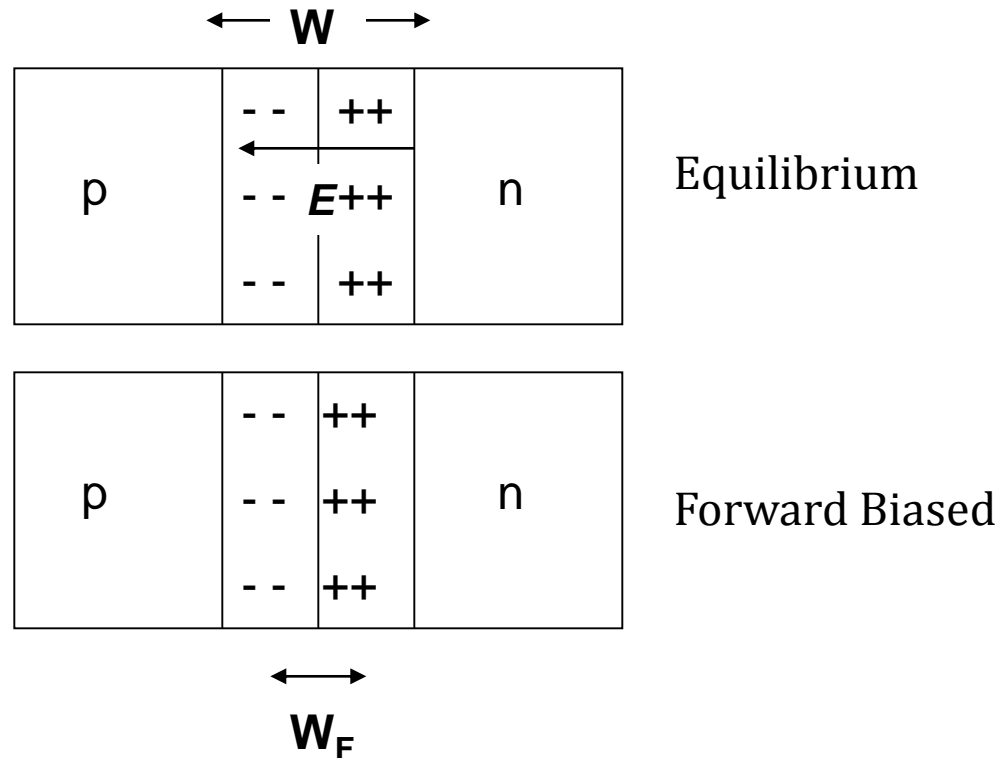


- **+ve terminal** is applied to the **p-region** of the pn junction and vice versa.
- Direction of the **applied electric field E_A** is the **opposite** as that of the E -field in the space-charge region.
- The **net result is that the electric field** in the space-charge region **lower** than the thermal equilibrium value causing diffusion of charges to begin again.
- The diffusion process continues as long as V_D is applied.
- Creating current in the pn junction, I_D





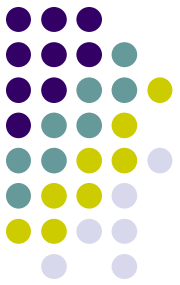
Forward-Biased pn Junction



Due to the diffusion of carriers \rightarrow the width of the depletion region reduces



Ideal Current-Voltage Relationship



Current I_D equation of a pn junction diode:

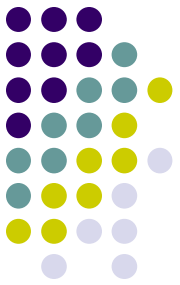
$$I_D = I_S \left[e^{\left(\frac{V_D}{n V_T} \right)} - 1 \right]$$

I_S = the reverse-bias saturation current (for silicon 10^{-15} to 10^{-13} A)

V_T = the thermal voltage (0.026 V at room temperature)

n = the emission coefficient ($1 \leq n \leq 2$) – normally it is set 1





Ideal Current-Voltage Relationship

Example

Determine the current in a pn junction diode.

Consider a pn junction at $T = 300$ K in which $I_S = 1.4 \times 10^{-14}$ A and $n = 1.1$

Find the diode current for $v_D = +0.75$ V and $v_D = -0.75$ V.

Solution: For $V_D = +0.75$ V, the diode is forward-biased and

$$I_D = I_S \left[e^{\left(\frac{V_D}{n V_T} \right)} - 1 \right] = (1.4 \times 10^{-14}) \left[e^{\left(\frac{0.75}{1.1 \times 0.026} \right)} - 1 \right] = 3.427 \text{ mA}$$

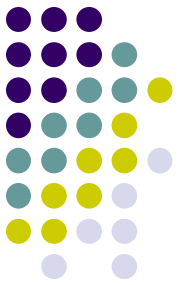
For $V_D = -0.75$ V, diode is reverse-biased and

$$I_D = I_S \left[e^{\left(\frac{V_D}{n V_T} \right)} - 1 \right] = (1.4 \times 10^{-14}) \left[e^{\left(\frac{-0.75}{1.1 \times 0.026} \right)} - 1 \right] = -1.4 \times 10^{-14} \text{ A}$$



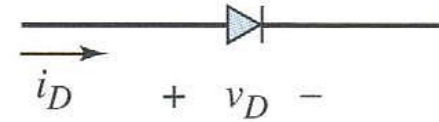
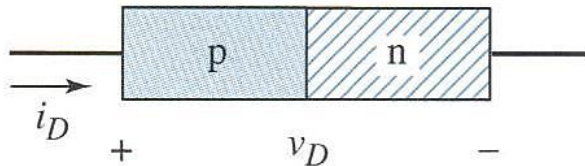
Very small current



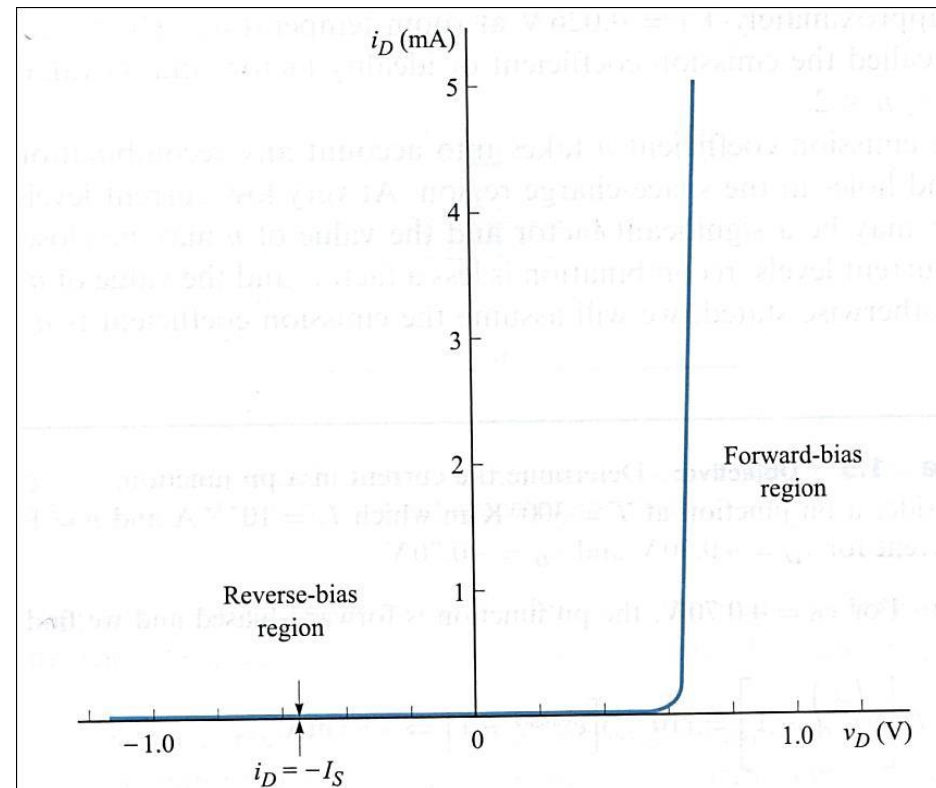


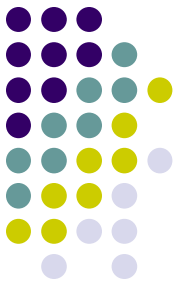
PN Junction Diode

- The basic PN junction diode circuit symbol, and conventional current direction and voltage polarity.



- The graphs shows the ideal I-V characteristics of a PN junction diode.
- The diode current is an exponential function of diode voltage in the forward-bias region.
- The current is very nearly zero in the reverse-bias region.





PN Junction Diode

- Temperature Effects

- Both I_S and V_T are functions of temperature.
- The diode characteristics vary with temperature.
- For silicon diodes, the change is approximately 2 mV/°C.

- Forward-biased PN junction characteristics versus temperature.
- The required diode voltage, V_γ to produce a given current **decreases** with an **increase in temperature**.

